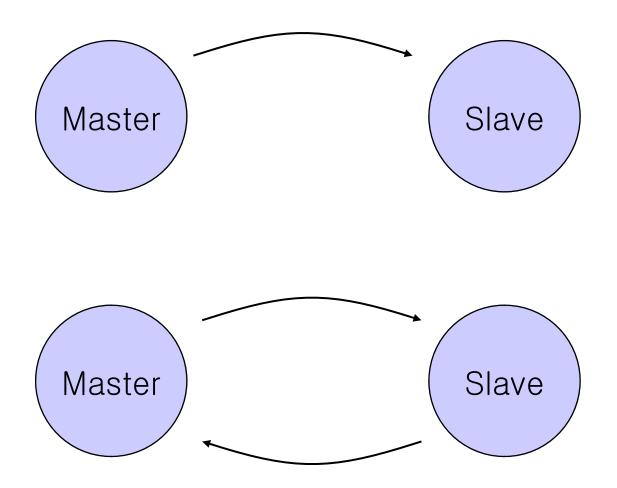
### Control of Telerobotic Systems

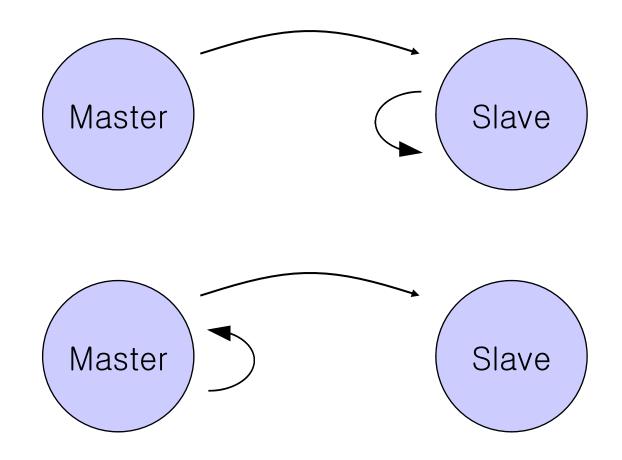
Jee-Hwan Ryu

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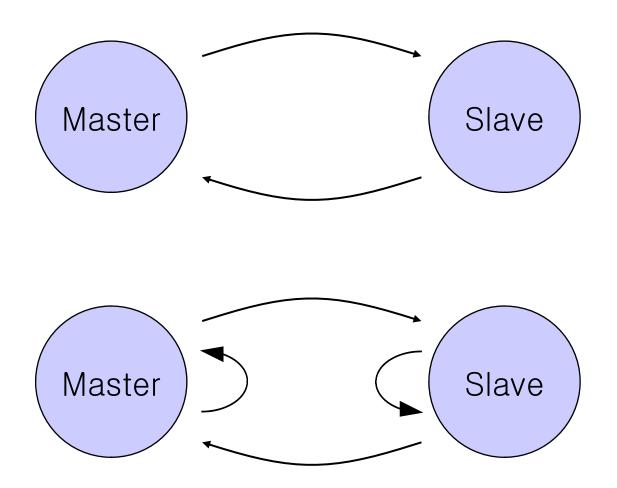
#### Unilateral vs. Bilateral



#### **Unilateral Control**



#### **Bilateral Control**



## Control Objectives of Bilateral Control

1. Ideal response

#### Two Aspects in Control of Teleoperator

#### Performance

 Make the operator feel as if he/she directly interact with the remote environment

#### Stability

Endure stable operation under wide variety of operating conditions

#### Ideal Teleoperator [Hannaford, 1989]

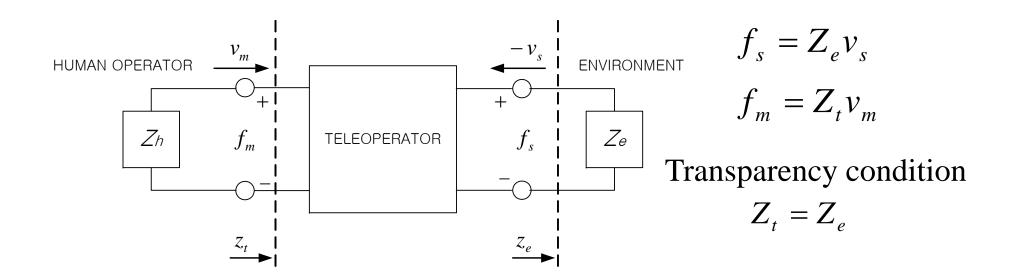
The hybrid two-port network model of teleoperator

Hybrid matrix

$$\begin{bmatrix} f_m \\ -v_s \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} v_m \\ f_s \end{bmatrix} \quad h = \begin{bmatrix} \text{Input impedance} & \text{Force scale} \\ -\text{Velocity scale} & \text{output admittance} \end{bmatrix}$$

$$h_{ideal} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$

### Transparency [Lawrence, 1993]



The essential desire is to provide a faithful transmission of signals (positions, velocities, forces) between master and slave to couple the operator as closely as possible to the remote task.

Ideally, the teleoperation system would be completely transparent, so operators feel that they are directly interacting with the remote task.

#### Ideal response: ideal kinesthetic coupling

[Yokokohji, 1994]

• *Ideal response I*: the position responses by the operator's input are identical, whatever the object dynamics is.

$$x_m = x_s$$

• *Ideal response II*: the force responses by the operator's input are identical, whatever the object dynamics is.

$$f_m = f_s$$

• *Ideal response III*: both the position responses and the force responses by the operator's input are identical respectively, whatever the object dynamics is.

$$x_m = x_s$$
 &  $f_m = f_s$ 

#### Arbitrarily position/force scaling [Ryu, 1999]

$$x_s = \lambda_p x_m$$
$$\lambda_f f_s = f_m$$

$$\lambda_f f_s = f_m$$

# Position/Force Matching vs. Impedance Matching

Position/Force matching

$$x_s = x_m$$

$$f_{s} = f_{m}$$

Impedance matching

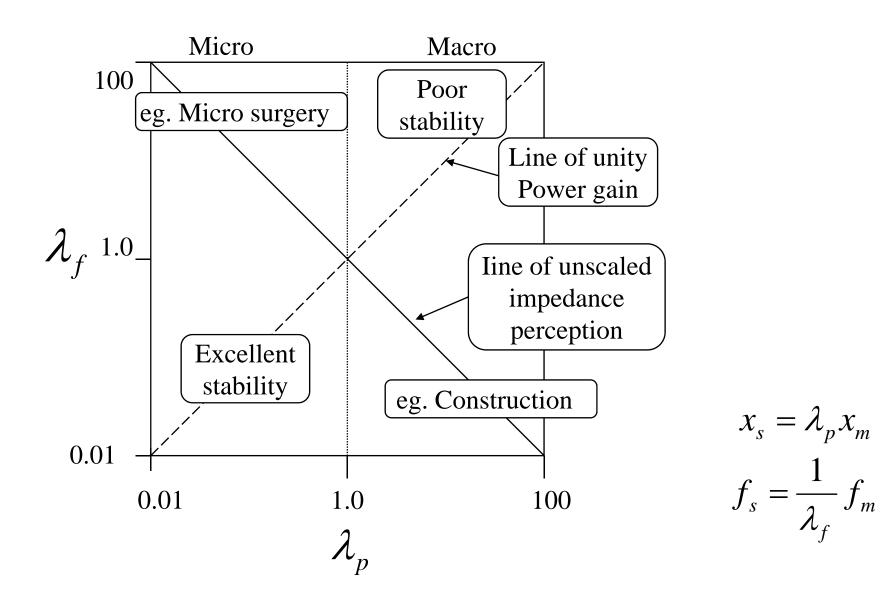
$$Z_t = Z_e$$

Position/Force matching → Impedance matching

Position/Force matching 

✓ Impedance matching

#### Characteristics on Scaling



#### **Architectures of Bilateral Control**

- 1. P/P
- 2. P/F
- 3. F/F
- 4. PF/PF
- 5. Local Force Feedback

#### General Bilateral Control Architecture

$$\tau_{m} = \begin{bmatrix} K_{mpm} + K'_{mpm} \frac{d}{dt} + K''_{mpm} \frac{d^{2}}{dt^{2}} & K_{mfm} \end{bmatrix} \begin{bmatrix} x_{m} \\ f_{m} \end{bmatrix}$$

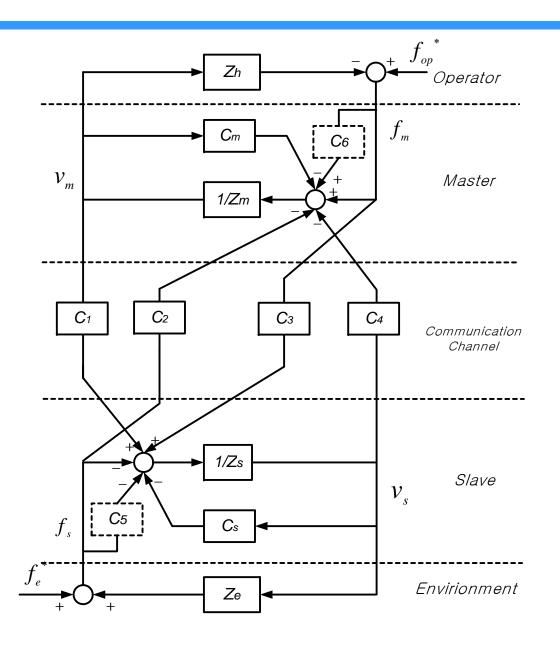
$$- \begin{bmatrix} K_{mps} + K'_{mps} \frac{d}{dt} + K''_{mps} \frac{d^{2}}{dt^{2}} & K_{mfs} \end{bmatrix} \begin{bmatrix} x_{s} \\ f_{s} \end{bmatrix}$$

$$\tau_{s} = \begin{bmatrix} K_{spm} + K'_{spm} \frac{d}{dt} + K''_{spm} \frac{d^{2}}{dt^{2}} & K_{sfm} \end{bmatrix} \begin{bmatrix} x_{m} \\ f_{m} \end{bmatrix}$$

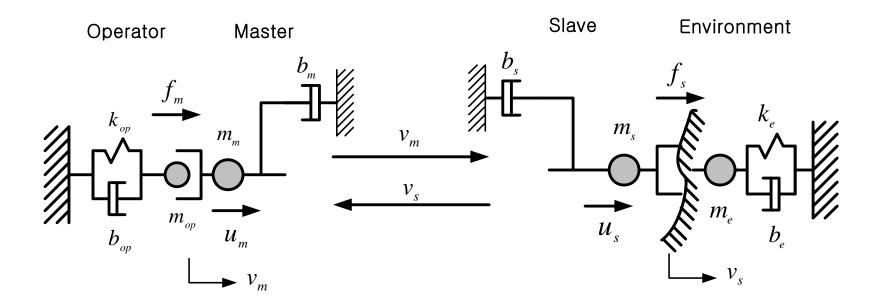
$$- \begin{bmatrix} K_{sps} + K'_{sps} \frac{d}{dt} + K''_{sps} \frac{d^{2}}{dt^{2}} & K_{sfs} \end{bmatrix} \begin{bmatrix} x_{s} \\ f_{s} \end{bmatrix}$$

Use all 4 information for control

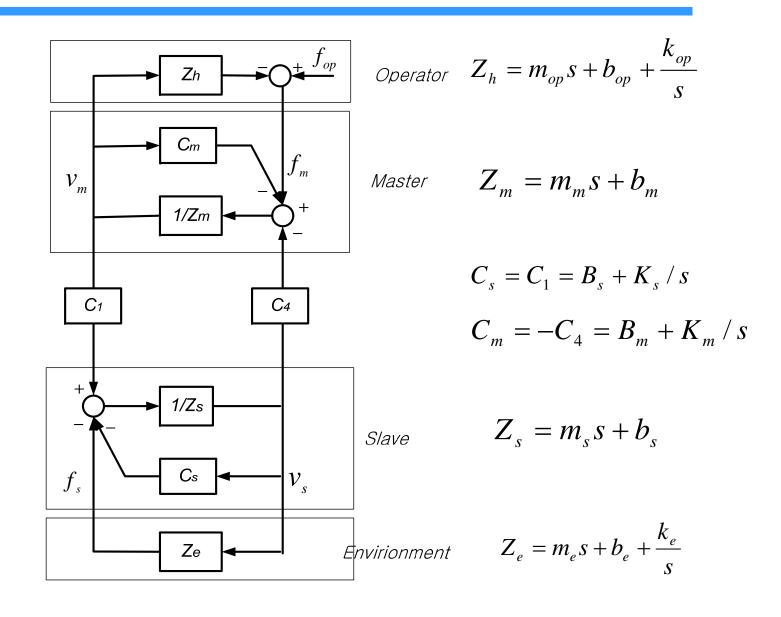
### General Structure 4-channel



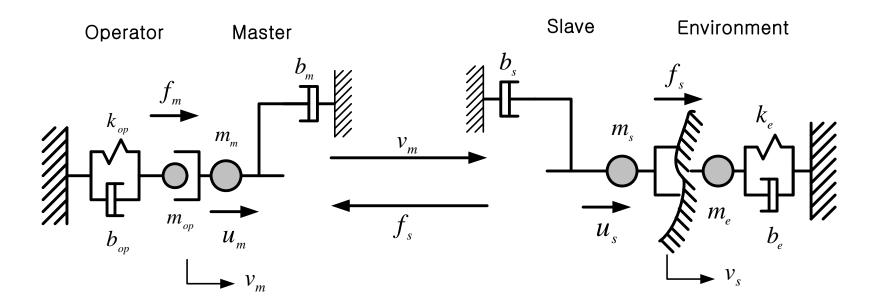
#### Position/Position Architecture



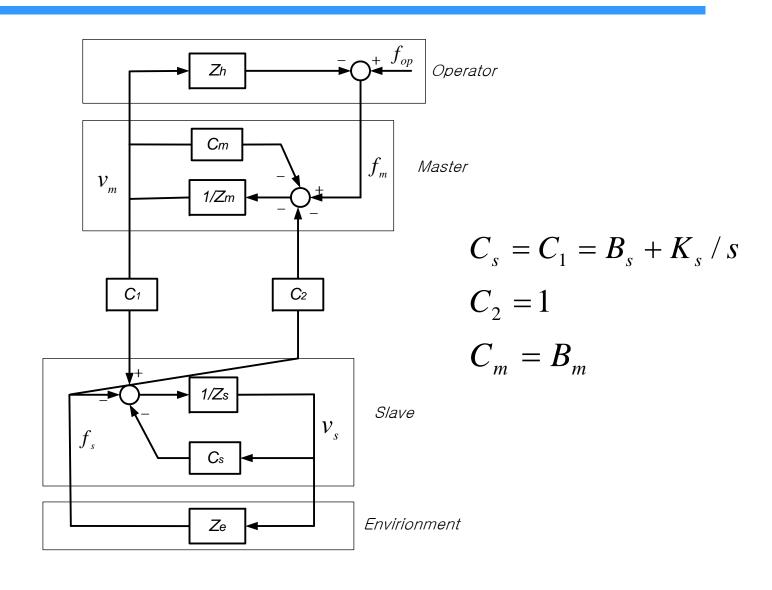
#### Position/Position Architecture



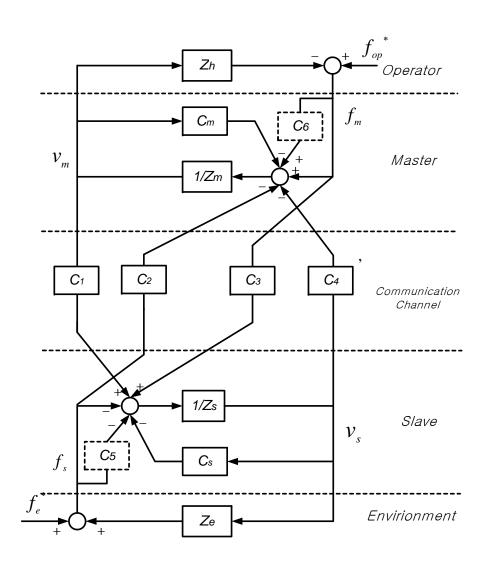
#### Position/Force Architecture



#### Position/Force Architecture



### General Structure (4-channel)



Position-position architecture :

$$C_1 \neq 0, C_4 \neq 0, C_2 = C_3 = 0$$
  
 $C_5 = C_6 = 0$ 

Position-force architecture:

$$C_1 \neq 0, C_2 \neq 0, C_3 = C_4 = 0$$
  
 $C_5 = C_6 = 0$ 

Force-position architecture:

$$C_3 \neq 0, C_4 \neq 0, C_1 = C_2 = 0$$
  
 $C_5 = C_6 = 0$ 

Force-force architecture:

$$C_2 \neq 0, C_3 \neq 0, C_1 = C_4 = 0$$
  
 $C_5 = C_6 = 0$ 

### Comparing of the Architectures

The transmitted impedance to the operator

$$Z_{t} = \frac{A + CZ_{e}}{B + DZ_{e}}$$

where

$$A = (Z_m + C_m)(Z_s + C_s) + C_1C_4$$

$$B = (1 + C_6)(Z_s + C_s) - C_3C_4$$

$$C = (1 + C_5)(Z_m + C_m) + C_1C_2$$

$$D = (1 + C_5)(1 + C_6) - C_2C_3$$

#### Position-Position Architecture

$$Z_{t} = \frac{(Z_{m} + C_{m})(Z_{s} + C_{s}) - C_{m}C_{s} + (Z_{m} + C_{m})Z_{e}}{(Z_{s} + C_{s}) + Z_{e}}$$

$$h = \begin{bmatrix} Z_m + C_m - \frac{C_m C_s}{Z_s + C_s} & \frac{C_m}{Z_s + C_s} \\ -\frac{C_s}{Z_s + C_s} & \frac{1}{Z_s + C_s} \end{bmatrix}$$

If  $C_m, C_s$  goes to infinite

$$h \Rightarrow \begin{bmatrix} Z_m + C_m - \frac{C_m C_s}{Z_s + C_s} & 1\\ -1 & 0 \end{bmatrix}$$

#### Position-Force Architecture

$$Z_{t} = \frac{(Z_{m} + C_{m})(Z_{s} + C_{s}) + (Z_{m} + C_{m} + C_{s})Z_{e}}{(Z_{s} + C_{s}) + Z_{e}}$$

$$h = \begin{bmatrix} Z_m + C_m & 1 \\ -C_s & 1 \\ \overline{Z_s + C_s} & \overline{Z_s + C_s} \end{bmatrix}$$

if 
$$C_s$$
 goes to infinite  $h \Rightarrow \begin{bmatrix} Z_m + C_m & 1 \\ -1 & 0 \end{bmatrix}$ 

If 
$$C_s$$
 goes to infinite and  $C_m = -Z_m$   $h \Rightarrow \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$ 

# Transparency optimized control law with four-channel

$$C_1 = (Z_s + C_s)$$
  $C_2 C_3 = 1$   
 $C_5 = C_6 = 0$   $C_4 = -(Z_m + C_m)$ 

$$\Rightarrow Z_t = Z_e \qquad h = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$

#### With local force feedback [H-Zaad, 1999]

$$C_{1} = (Z_{s} + C_{s}) \qquad C_{2} = 1 + C_{6}$$

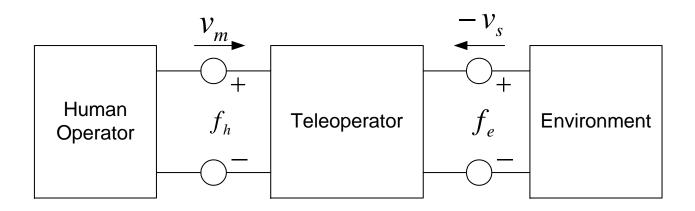
$$C_{3} = 1 + C_{5} \qquad C_{4} = -(Z_{m} + C_{m})$$

$$\Rightarrow \qquad Z_{t} = Z_{e} \qquad h = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$

Increase stability margin for the time delay teleoperation, because the feed-forward control gain  $C_2, C_3$  can be attenuated by the local force feedback gain  $C_5, C_6$ 

However, acceleration should be measured and the dynamic parameters of the master and slave should be known, perfectly.

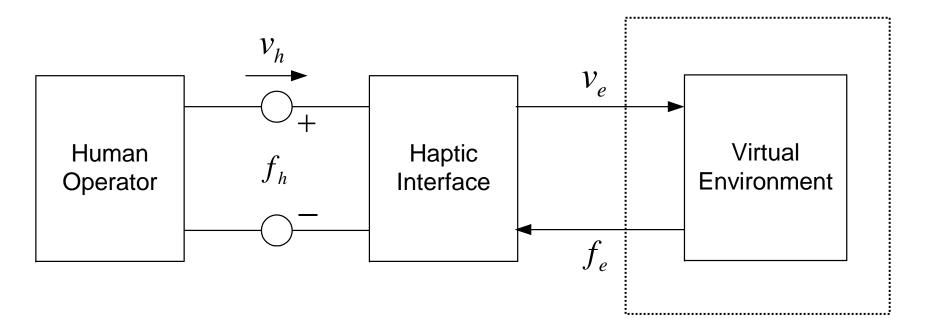
# Network Model and Stability Condition



Teleoperator two-port should be passive

$$\int_0^t (f_h(\tau)v_m(\tau) - f_e(\tau)v_s(\tau))d\tau \ge 0, \qquad \forall t \ge 0$$

# Network Model and Stability Condition



Virtual Environment one-port should be passive

$$\int_0^t f_e(\tau) v_e(\tau) d\tau \ge 0, \qquad \forall t \ge 0$$

### **Passivity**

- Principle of conservation of energy:
  - "Energy supplied BY the network can never exceed the energy which has been fed TO it"
- Mathematical definitions

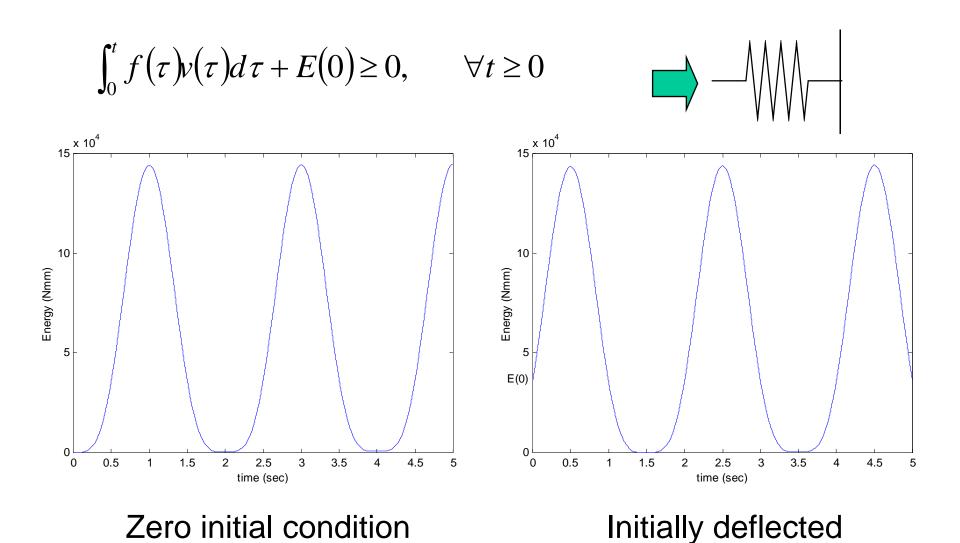
$$\int_0^t f(\tau)v(\tau)d\tau + E(0) \ge 0, \qquad \forall t \ge 0$$

Net energy supplied Initial energy storage

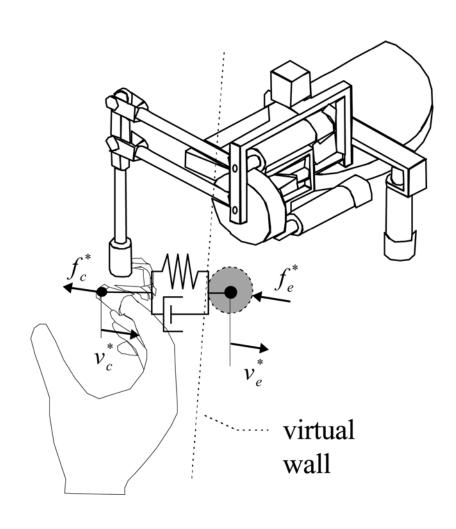
 $f \cdot v > 0 \qquad \qquad \uparrow \qquad \qquad \uparrow \qquad \qquad f \qquad \qquad f \qquad \qquad f \qquad \qquad \uparrow \qquad \qquad \uparrow \qquad \qquad \uparrow \qquad \qquad \downarrow \qquad \qquad \uparrow \qquad \qquad \downarrow \qquad \qquad \downarrow$ f: Force

v: Velocity

## **Energy Behavior of Spring**



# Virtual Coupling



# Passivity Observer (PO) can measure energy flow in real-time

#### Passivity:

$$\int_0^t f(\tau)v(\tau)d\tau \ge 0, \qquad \forall t \ge 0$$

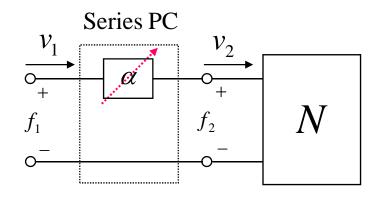
PO: 
$$E_{obsv}(n) = \Delta T \sum_{k=0}^{n} f(k)v(k)$$

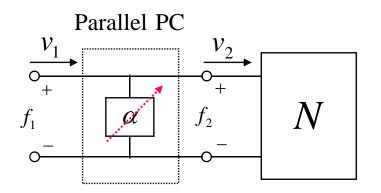
$$E_{obsv}(n) \ge 0 : \text{Passive}$$

$$E_{obsv}(n) < 0 : \text{Active}$$

-Hannaford and Ryu 2001-

# Passivity Controller (PC) is an adaptive dissipation element





Series or velocity conserving

Impedance causality

parallel or force conserving

Admittance causality

-Hannaford and Ryu 2001-

### Series PC Algorithm

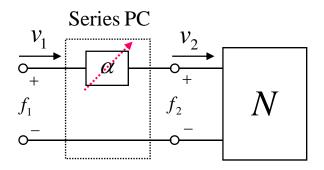
- 1)  $v_1(n) = v_2(n)$  is an input
- 2)  $f_2(n) = F_N(v_2(n))$

where  $F_N()$  is the output of the one-port

3) 
$$E_{obsv}(n) = E_{obsv}(n-1) + [f_2(n)v_2(n) + \alpha(n-1)v_2(n-1)^2]\Delta T$$

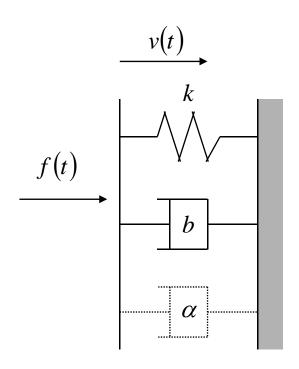
4) 
$$\alpha(n) = \begin{cases} -E_{obsv}(n)/\Delta T \ v_2(n)^2 & \text{if } E_{obsv}(n) < 0 \\ 0 & E_{obsv}(n) \ge 0 \end{cases}$$

5) 
$$f_1(n) = f_2(n) + \alpha(n) v_2(n) \Rightarrow \text{output}$$



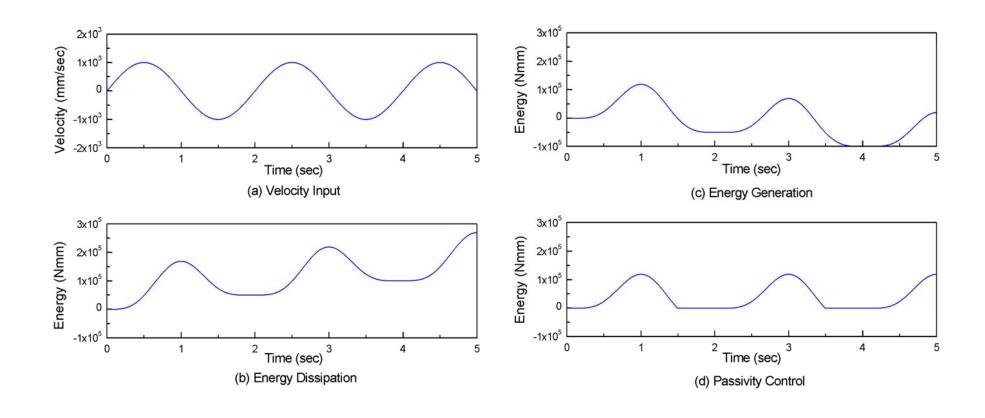
-Hannaford and Ryu 2001-

# Simple Simulation with Impedance Type Virtual Wall

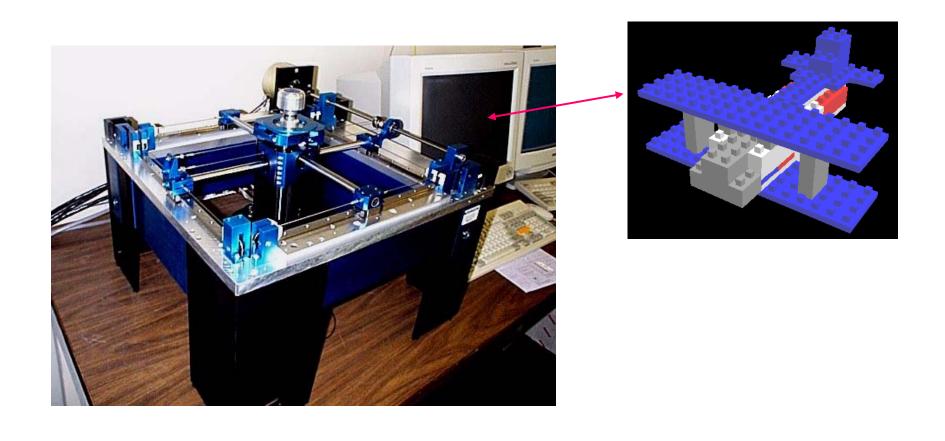


$$k = 710 \ N/m$$
  
 $b = 50 \ Ns/m$ 

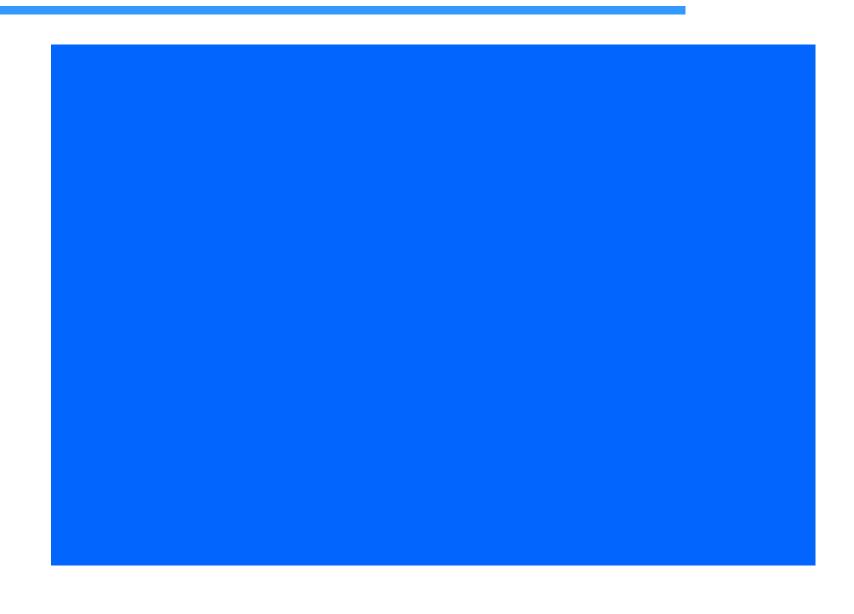
#### Simulation Results



# Excalibur Haptic Interface System



# Haptic Experiment with the PC



## Teleoperation Experiment with the PC

